General Description

The ICS840N051I is a LVCMOS/LVTTL clock synthesizer designed for SDH/SONET and Ethernet applications. The device generates a selectable 155.52MHz or 77.76MHz clock signal with excellent phase jitter performance. The device uses IDT's fourth generation FemtoClock® NG technology for an optimum of high clock frequency, low phase noise performance and low power consumption. The device supports 2.5V or 3.3V voltage supply and is packaged in a small, lead-free (RoHS 6) 8-lead TSSOP package. The extended temperature range supports wireless infrastructure, telecommunication and networking end equipment requirements.

Features

- Fourth generation FemtoClock[®] NG technology
- 155.52MHz output clock synthesized from a 19.44MHz fundamental mode crystal
- One 2.5V or 3.3V LVCMOS/LVTTL clock output
- Crystal interface designed for a 12pF parallel resonant crystal
- RMS phase jitter @ 155.52MHz, using a 19.44MHz crystal (12kHz - 20MHz): 0.482ps (maximum)
- RMS phase jitter @ 156.25MHz, using a 19.53125MHz crystal (1.875MHz - 20MHz): 0.138ps (maximum)
- LVCMOS interface levels for the control inputs
- Full 2.5V or 3.3V supply voltage
- Lead-free (RoHS 6) packaging
- -40°C to 85°C ambient operating temperature

OE Function Table

Input	
OE	Output Enable
0	Output Q is disabled in high-impedance state
1 (default)	Output Q is enabled.

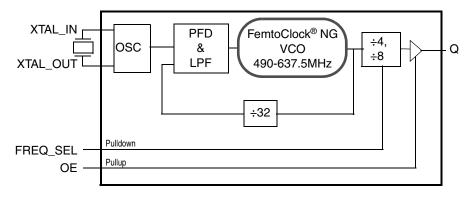
NOTE: OE is an asynchronous control

FREQ_SEL Frequency Table

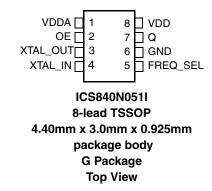
Input	Output Frequency			
FREQ_SEL	f _{XTAL} = 19.2MHz	f _{XTAL} = 19.44MHz	f _{XTAL} = 19.53125MHz	
0 (default)	153.6MHz	155.52MHz	156.25MHz	
1	76.8MHz	77.76MHz	78.125MHz	

NOTE: FREQ_SEL is an asynchronous control.

Block Diagram



Pin Assignment



Pin Descriptions and Characteristics

Table 1. Pin Descriptions

Number	Name	Туре		Description
1	V_{DDA}	Power		Analog power supply.
2	OE	Input	Pullup	Output enable pin. LVCMOS interface levels.
3, 4	XTAL_OUT, XTAL_IN	Input		Crystal oscillator interface. XTAL_IN is the input, XTAL_OUT is the output.
5	FREQ_SEL	Input	Pulldown	Frequency select pin. LVCMOS interface levels.
6	GND	Power		Power supply ground.
7	Q	Output		Single-ended clock output. LVCMOS/LVTTL interface levels.
8	V_{DD}	Power		Core supply pin.

NOTE: Pulldown and Pullup refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

Table 2. Pin Characteristics

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C _{IN}	Input Capacitance	OE, FREQ_SEL		3.5		pF
C	Power Dissipation	V _{DD} = 3.465V		11		pF
Capacitance	V _{DD} = 2.625V		9		pF	
R _{Pullup}	Input Pullup Resistor			51		kΩ
R _{Pulldown}	Input Pulldown Resistor			51		kΩ
R _{OUT}	Output Impadance	V _{DD} = 3.3V		15		Ω
	Output Impedance	V _{DD} = 2.5V		19		Ω

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, V _{DD}	3.63V
Inputs, V _I XTAL_IN Other Inputs	0V to 2V -0.5V to V _{DD} + 0.5V
Outputs, V _O	-0.5V to V _{DD} + 0.5V
Package Thermal Impedance, θ_{JA}	117°C/W (0 mps)
Storage Temperature, T _{STG}	-65°C to 150°C

DC Electrical Characteristics

Table 3A. Power Supply DC Characteristics, $V_{DD} = 3.3V \pm 5\%$ or $2.5V \pm 5\%$, $T_A = -40$ °C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{DD}	Core Supply Voltage		2.375	3.3	3.465	V
V_{DDA}	Analog Supply Voltage		V _{DD} – 0.18	3.3	V _{DD}	V
V_{DDA}	Analog Supply Voltage		V _{DD} – 0.18	2.5	V _{DD}	٧
I _{DDA}	Analog Supply Current				18	mA
I _{DD}	Power Supply Current				67	mA

Table 3B. LVCMOS/LVTTL DC Characteristics, $V_{DD} = 3.3V \pm 5\%$ or $2.5V \pm 5\%$, $T_A = -40$ °C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V	Input High Volta		V _{DD} = 3.3V	2		V _{DD} + 0.3	V
V _{IH}	Input High Volta	ige	V _{DD} = 2.5V	1.7		V _{DD} + 0.3	V
		FREQ_SEL	V _{DD} = 3.3V	-0.3		0.5	V
\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	Input	OE	V _{DD} = 3.3V	-0.3		0.8	V
V _{IL}	Low Voltage	FREQ_SEL	V _{DD} = 2.5V	-0.3		0.5	V
		OE	V _{DD} = 2.5V	-0.3		0.7	V
	Input	FREQ_SEL	V _{DD} = V _{IN} = 3.465V or 2.625V			150	μA
IIH	High Current	OE	$V_{DD} = V_{IN} = 3.465V \text{ or } 2.625V$			5	μΑ
	Input	FREQ_SEL	V _{DD} = 3.465V or 2.625V, V _{IN} = 0V	-5			μΑ
I _{IL}	Low Current	OE	V _{DD} = 3.465V or 2.625V, V _{IN} = 0V	-150			μΑ
	Output	_	V _{DD} = 3.465V	2.6			V
V _{OH}	High Voltage; NOTE 1	Q	V _{DD} = 2.625V	1.8			V
V _{OL}	Output Low Voltage; NOTE 1	Q	V _{DD} = 3.465V or 2.625V			0.5	V

NOTE 1: Output terminated with 50Ω to V_{DD} / 2. See Parameter Measurement Information Section, LVCMOS Output Load Test Circuit Diagrams.

Table 4. Crystal Characteristics

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation			Fundamental		
Frequency		15.31	19.44	19.92	MHz
Equivalent Series Resistance (ESR)				80	Ω
Shunt Capacitance				7	pF
Capacitive Load (CL)			12		pF

AC Characteristics

Table 5. AC Characteristics, $V_{DD} = V_{DDA} = 3.3V \pm 5\%$ or $2.5V \pm 5\%$, $T_A = -40^{\circ}C$ to $85^{\circ}C$

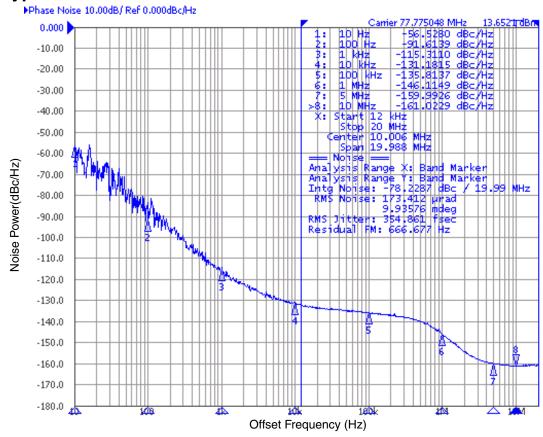
Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
ı	Output Fraguency	FREQ_SEL = 0	122.5	155.52	159.38	MHz
f _{OUT}	Output Frequency	FREQ_SEL = 1	61.25	77.76	79.69	MHz
		f _{OUT} = 155.52MHz, Integration Range: 12kHz – 20MHz, 19.44MHz crystal		0.350	0.482	ps
<i>t</i> jit(Ø)	RMS Phase Jitter (Random); NOTE 1	f _{OUT} = 77.76MHz, Integration Range: 12kHz – 20MHz, 19.44MHz crystal		0.354	0.508	ps
		f _{OUT} = 156.25MHz, Integration Range: 1.875MHz – 20MHz, 19.353125MHz crystal		0.101	0.138	ps
		f _{OUT} = 156.25MHz, Offset: 10Hz		-43.6		dBc/Hz
		f _{OUT} = 156.25MHz, Offset: 100Hz		-74.1		dBc/Hz
	0: 1 0:1 5 111:	f _{OUT} = 156.25MHz, Offset: 1kHz		-107.3		dBc/Hz
Φ_{N}	Single-Side Band Noise Power	f _{OUT} = 156.25MHz, Offset: 10kHz		-124.4		dBc/Hz
	1 GWGI	f _{OUT} = 156.25MHz, Offset: 100kHz		-128.9		dBc/Hz
		f _{OUT} = 156.25MHz, Offset: 1MHz		-139.1		dBc/Hz
		f _{OUT} = 156.25MHz, Offset: 10MHz		-156.7		dBc/Hz
t_R / t_F	Output Rise/Fall Time	20% to 80%	200		600	ps
odc	Output Duty Cycle		48		52	%

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

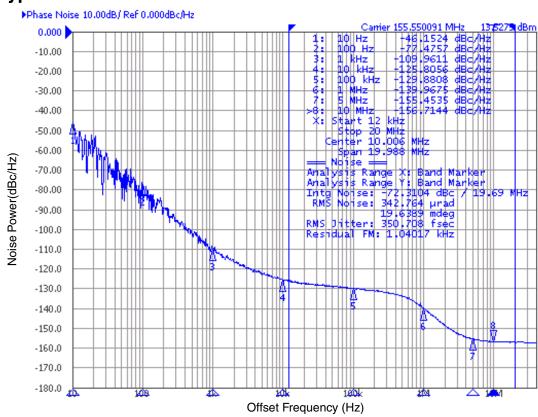
NOTE: Characterized with 19.2MHz, 19.44MHz and 19.53125MHz crystals.

NOTE 1: Please refer to the phase noise plots.

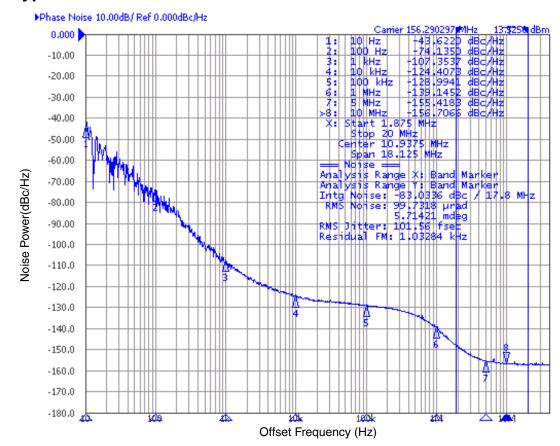
Typical Phase Noise at 77.76MHz



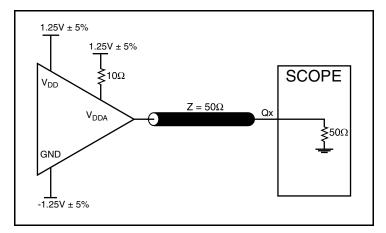
Typical Phase Noise at 155.52MHz



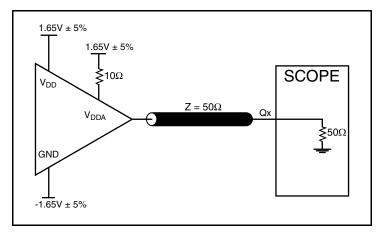
Typical Phase Noise at 156.25MHz



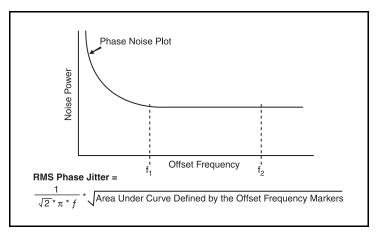
Parameter Measurement Information



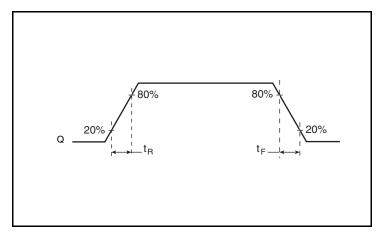
2.5V LVCMOS/LVTTL Output Load AC Test Circuit



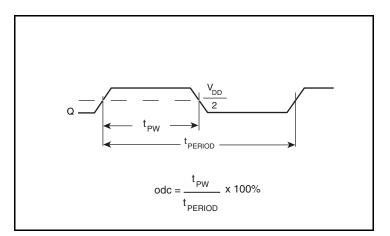
3.3V LVCMOS/LVTTL Output Load AC Test Circuit



RMS Phase Jitter



Output Rise/Fall Time



Output Duty Cycle/Pulse Width/Period

Applications Information

Overdriving the XTAL Interface

The XTAL_IN input can be overdriven by an LVCMOS driver or by one side of a differential driver through an AC coupling capacitor. The XTAL_OUT pin can be left floating. The amplitude of the input signal should be between 500mV and 1.8V and the slew rate should not be less than 0.2V/nS. For 3.3V LVCMOS inputs, the amplitude must be reduced from full swing to at least half the swing in order to prevent signal interference with the power rail and to reduce internal noise. Figure 1A shows an example of the interface diagram for a high speed 3.3V LVCMOS driver. This configuration requires that the sum of the output impedance of the driver (Ro) and the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This

can be done in one of two ways. First, R1 and R2 in parallel should equal the transmission line impedance. For most 50Ω applications, R1 and R2 can be 100Ω . This can also be accomplished by removing R1 and changing R2 to 50Ω . The values of the resistors can be increased to reduce the loading for a slower and weaker LVCMOS driver. Figure 1B shows an example of the interface diagram for an LVPECL driver. This is a standard LVPECL termination with one side of the driver feeding the XTAL_IN input. It is recommended that all components in the schematics be placed in the layout. Though some components might not be used, they can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a quartz crystal as the input.

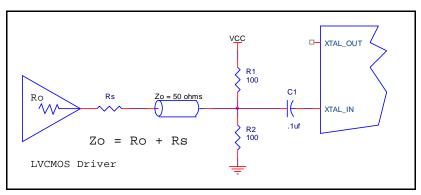


Figure 1A. General Diagram for LVCMOS Driver to XTAL Input Interface

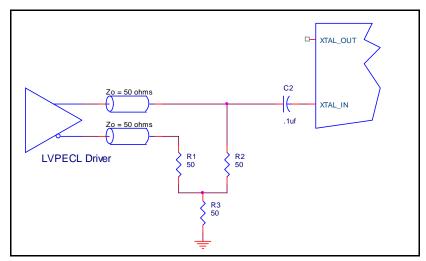


Figure 1B. General Diagram for LVPECL Driver to XTAL Input Interface

Schematic Layout

Figure 2 shows an example ICS840N051I application schematic in which the device is operated at $V_{DD} = V_{DDA} = 3.3V$. The schematic example focuses on functional connections and is intended as an example only and may not represent the exact user configuration. Refer to the pin description and functional tables in the datasheet to ensure the logic control inputs are properly set. For example OE and FREQ_SEL can be configured from an FPGA instead of set with pull up and pull down resistors as shown.

As with any high speed analog circuitry, the power supply pins are vulnerable to random noise, so to achieve optimum jitter performance isolation of the V_{DD} pin from power supply is required. In order to achieve the best possible filtering, it is recommended that the placement of the filter components be on the device side of the PCB

as close to the power pins as possible. If space is limited, the $0.1\mu F$ capacitor on the V_{DD} pin must be placed on the device side with direct return to the ground plane though vias. The remaining filter components can be on the opposite side of the PCB.

Power supply filter recommendations are a general guideline to be used for reducing external noise from coupling into the devices. The filter performance is designed for wide range of noise frequencies. This low-pass filter starts to attenuate noise at approximately 10kHz. If a specific frequency noise component is known, such as switching power supply frequencies, it is recommended that component values be adjusted and if required, additional filtering be added. Additionally, good general design practices for power plane voltage stability suggests adding bulk capacitances in the local area of all devices.

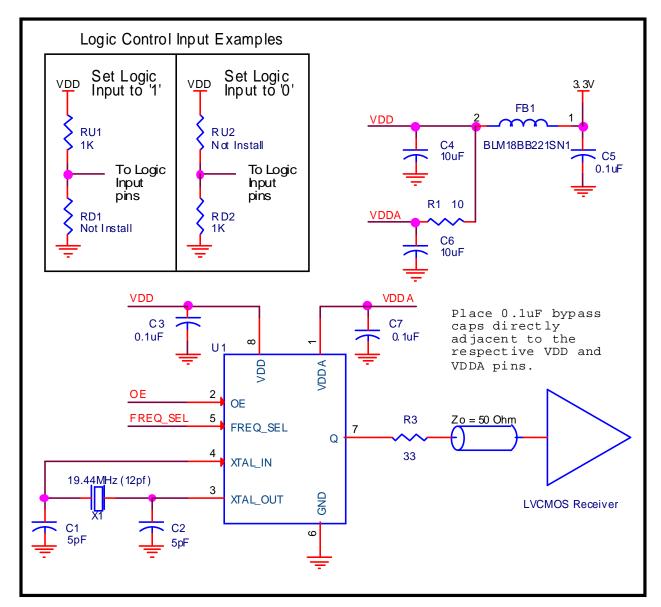


Figure 2. ICS840N051I Application Schematic

Power Considerations

This section provides information on power dissipation and junction temperature for the ICS840N051I. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS840N051I is the sum of the core power plus the analog power plus the power dissipated into the load. The following is the power dissipation for $V_{DD} = 3.3V + 5\% = 3.465V$, which gives worst case results.

- Power (core)_{MAX} = V_{DD MAX} * (I_{DD} + I_{DDA}) = 3.465V *(67mA + 18mA) = 294.53mW
- Output Impedance R_{OUT} Current due to Loading 50Ω to $V_{DD}/2$ Output Current $I_{OUT} = V_{DD~MAX} / [2*(50\Omega + R_{OUT})] = 3.465 V / [2*(50\Omega + 15\Omega)] = 26.7mA$
- Power Dissipation on the R_{OUT} per LVCMOS output Power (R_{OUT}) = R_{OUT} * (I_{OUT})² = 15 Ω * (26.7mA)² = **10.7mW per output**
- Total Power (R_{OUT}) = 10.7mW * 1 = 10.7mW

Dynamic Power Dissipation at 156.25MHz

```
Power (156.25MHz) = C_{PD} * Frequency * (V_{DD})^2 = 11pF * 156.25MHz * (3.465V)^2 = 20.64mW per output Total Power (156.25MHz) = 20.64mW * 1 = 20.64mW
```

Total Power Dissipation

- Total Power
 - = Power (core)_{MAX} + Power (R_{OUT}) + Power (156.25MHz)
 - = 294.53mW + 10.7mW + 20.64mW
 - = 325.87mW

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj = θ_{JA} * Pd_total + T_A

Tj = Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance q_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 117°C/W per Table 6 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 $85^{\circ}\text{C} + 0.326\text{W} * 117^{\circ}\text{C/W} = 123.2^{\circ}\text{C}$. This is below the limit of 125°C .

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 6. Thermal Resistance θ_{JA} for 8 Lead TSSOP, Forced Convection

θ_{JA} by Velocity			
Meters per Second	0		
Multi-Layer PCB, JEDEC Standard Test Boards	117°C/W		

Reliability Information

Table 7. $\theta_{\mbox{\scriptsize JA}}$ vs. Air Flow Table for a 8-lead TSSOP

θ_{JA} vs. Air Flow			
Meters per Second	0		
Multi-Layer PCB, JEDEC Standard Test Boards	117°C/W		

Transistor Count

The transistor count for ICS840N051I is: 24,811

Package Outline and Package Dimensions

Package Outline - G Suffix for 8 Lead TSSOP

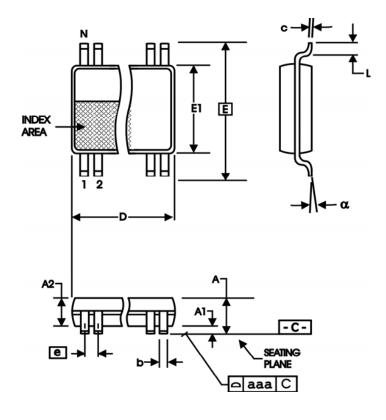


Table 8. Package Dimensions

All Dimensions in Millimeters				
Symbol	Minimum	Maximum		
N	8	3		
Α		1.20		
A1	0.5	0.15		
A2	0.80	1.05		
b	0.19 0.30			
С	0.09	0.20		
D	2.90	3.10		
Е	6.40	Basic		
E1	4.30	4.50		
е	0.65	Basic		
L	0.45	0.75		
α	0° 8°			
aaa		0.10		

Reference Document: JEDEC Publication 95, MO-153

Ordering Information

Table 9. Ordering Information

Part/Order Number	Marking	Package	Shipping Packaging	Temperature	
840N051BGILF	51BIL	Lead-Free, 8-lead TSSOP	Tube	-40°C to 85°C	
840N051BGILFT	51BIL	Lead-Free, 8-lead TSSOP	Tape & Reel	-40°C to 85°C	

Revision History Sheet

Rev	Table	Page	Description of Change	Date
А		1	General Description - corrected output frequency of 156.25MHz to 155.52MHz in second sentence.	10/14/2013

We've Got Your Timing Solution



6024 Silver Creek Valley Road San Jose, California 95138 Sales

800-345-7015 (inside USA) +408-284-8200 (outside USA) Fax: 408-284-2775 www.IDT.com/go/contactIDT

cantly affect the health or safety of users. Anyone using an IDT product in such a manner does so at their own risk, absent an express, written agreement by IDT.

Technical Support Sales

netcom@idt.com +480-763-2056

DISCLAIMER Integrated Device Technology, Inc. (IDT) and its subsidiaries reserve the right to modify the products and/or specifications described herein at any time and at IDT's sole discretion. All information in this document, including descriptions of product features and performance, is subject to change without notice. Performance specifications and the operating parameters of the described products are determined in the independent state and are not guaranteed to perform the same way when installed in customer products. The information contained herein is provided without representation or warranty of any kind, whether express or implied, including, but not limited to, the suitability of IDT's products for any particular purpose, an implied warranty of merchantability, or non-infringement of the intellectual property rights of others. This document is presented only as a guide and does not convey any license under intellectual property rights of IDT's products for any particular purpose.

IDT's products are not intended for use in applications involving extreme environmental conditions or in life support systems or similar devices where the failure or malfunction of an IDT product can be reasonably expected to signifi-

Integrated Device Technology, IDT and the IDT logo are registered trademarks of IDT. Other trademarks and service marks used herein, including protected names, logos and designs, are the property of IDT or their respective third protected names.

Copyright 2013. All rights reserved.